

DISCRETE COSINE TRANSFORM-II FOR REDUCTION IN PEAK TO AVERAGE POWER RATIO OF OFDM SIGNALS THROUGH μ -LAW COMPANDING TECHNIQUE

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Abstract

Orthogonal frequency Division multiplexing (OFDM) is the most familiar word in telecommunication and wireless communication systems as it provides enhanced spectral efficiency than Frequency division multiplexing (FDM). Although it is sustaining an orthogonal relation between carriers but high peak to average power ratio (PAPR) is one of the main disadvantages of OFDM system. Various PAPR reduction techniques have been used, including techniques based on companding. Incompanding, μ -Law companding has potential to reduce the PAPR of OFDM signals. μ -Law Companding technique preserves the dynamic range of samples at low amplitudes. A new method named as precoding which is having less complexity compared to the other power reduction techniques is proposed to reduce PAPR. This paper put forward combination of two existing techniques namely μ -Law Companding Transform and Discrete Cosine Transform-II precoding technique. The simulation results show that, the proposed combined scheme gives better result for PAPR Reduction and results in no distortion.

Keywords

FDM, OFDM, PAPR, FFT, IEEE, DCT, CCDF

1. INTRODUCTION

As the demand for high data rate wireless digital communication is rapidly expanding, several approaches such as increasing modulation order or employing multiple antennas at both transmitter and receiver that are reliable for wireless systems have been studied to enhance the spectral efficiency [1,2]. In today's communication systems Orthogonal Frequency Division Multiplexing (OFDM), widespread modulation technique has been considered to achieve high rate data transmission for wireless environment. OFDM system has high spectral efficiency, ease of implementation using the fast Fourier transform (FFT) and simple equalization techniques and also provides immunity to multipath fading, robustness against inter-symbol

interference. Therefore it is considered for digital communication systems and also for future broadband applications.

Multicarrier networks such as Frequency Division Multiplexing (FDM) have been around since the late 1950's [3], however due to their implementation complexity transmitted in parallel. Multicarrier systems in wireless applications are less susceptible to channel induced distortions than single carrier systems at corresponding data rates and inefficient use of the frequency band they were restricted to military applications.

In multicarrier transmission, a single data stream bearing a large number of narrow-band subcarriers is transmitted over a number of lower rate subcarriers. These subcarriers are regularly spaced in frequency, domain to avoid the interference. The subcarriers are totally independent of each other because the frequency spacing and time synchronization of the subcarriers is chosen in such a way that the subcarriers are orthogonal. By selecting a special set of orthogonal carrier frequencies, high spectral efficiency is obtained because the spectra of the sub-carriers do not overlap. The orthogonality among subcarriers can be maintained by introducing a cyclic prefix over a dispersive channel. The advantage of this structure over single carrier systems is that the extended symbol time due to lower data rate makes the signal less susceptible to effects of the channel such as multipath propagation which introduces Inter Symbol Interference (ISI). Each sub-channel will therefore experience flat fading reducing the equalization complexity in the receiver.

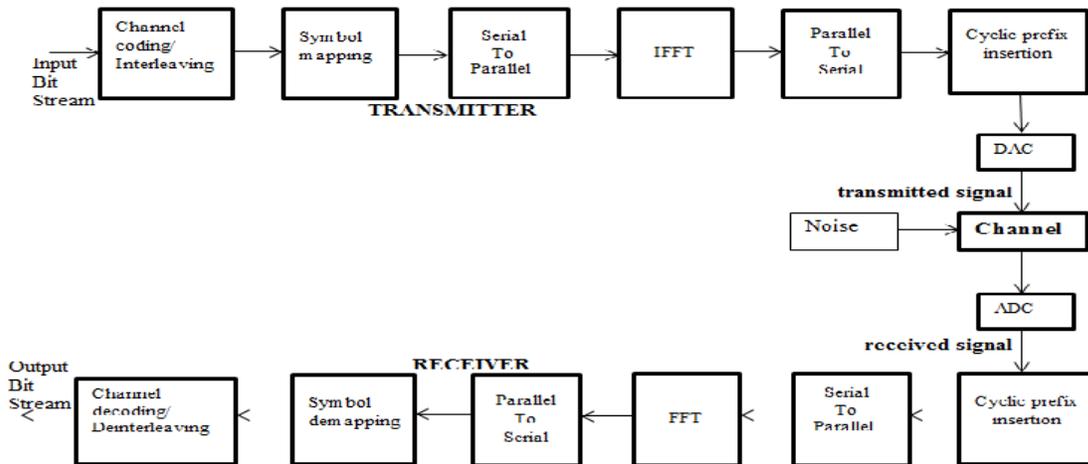


Fig. 1: Block Diagram of OFDM system

In an OFDM scheme shown in fig. 1, the orthogonal and overlapped subcarriers are a large number of orthogonal, overlapping, narrow band sub-carriers are transmitted in parallel. Maintaining total data rates similar to conventional single-carrier modulation schemes in the same bandwidth to modulate each sub-carrier is with a conventional modulation scheme such as Quadrature Amplitude Modulation (QAM) or Phase-Shift Keying (PSK) at a low symbol rate than that required for the whole data stream, but still. In OFDM system a high speed data is splits into a number of parallel lower data rate streams and these low rates data streams are transmitted simultaneously over a number of orthogonal subcarriers. The amount of dispersion in time caused due to multipath delay is reduced as, the symbol duration increases for lower rate parallel sub-carriers. These carriers divide the available transmission bandwidth.

The sub-carriers are separated in such a way so that there is a very compact spectral utilization and each being modulated at a low bit rate. In a conventional frequency division multiplex the carriers are spaced apart to avoid spectral overlap and multipath distortion.

OFDM has progressed to the point where it has now been widely used for various communication applications such as digital audio and video broadcasting, asymmetric digital subscriber line modems, and wireless local area networks systems(WLAN).It has also been adopted as the physical layer modulation scheme for wireless networking standards such as Hiperlan2 in Europe and the Institute of Electrical and Electronic Engineers (IEEE) 802.11a, g standards in the United States.

OFDM systems have a large dynamic signal range with a very high peak to average power ratio (PAPR) which is a one of major drawbacks of the transmitted OFDM signal. As a result, the OFDM signal will be clipped when passed through a nonlinear high power amplifier (HPA) at the transmitter as it only operate in its linear region with sufficient back-off. Thus, HPA with a large dynamic range is required for OFDM systems. These amplifiers are very expensive and are major cost component of the OFDM systems. Thus reduction in the value of the PAPR not only reduces the cost of OFDM systems and reduces the complexity of analog-to-digital (A/D) and digital-to-analog (D/A) converters, but also increases the transmit power, thus, improving received signal-noise-ratio (SNR).

To reduce the PAPR, several techniques have been proposed, which can be divided into three groups.The first group of reduction techniques includes signal distortion techniques, which reduce the peak amplitudes simply by nonlinearly distorting the OFDM signal at or around the peaks. Examples of distortion techniques are clipping [5], peak windowing [6], and peak cancellation. Second, there are coding techniques that use a special code set that excludes OFDM symbols with a large PAPR. The special codes includes Golay complementary sequence [7], Reed Muller code [8], and Hadamard code in this approach [9].The third technique scrambles each OFDM symbol with different scrambling sequences and selecting the sequence that gives the smallest PAPR such as selected mapping (SLM) [10] and partial transmit sequence (PTS) [11].

The organization of this paper is as follow. Section 2 presents formulates the problem of PAPR in OFDM system.The reduction technique based on μ -Law companding is explained in section 3.Section 4 describes the precoding based Discrete Cosine Transform-II technique. PAPR schemes based on combined DCT-II and μ -Law companding technique are proposed in section 5. Computer simulations results are presented in Section 6. Section 6 draws conclusions regarding the existing and proposed combined technique.

2. PAPR IN OFDM SIGNAL

Presence of large number of independently modulated sub-carriers in an OFDM system increases the peak value of the system as compared to the average of the whole system. The ratio of the peak to average power value is termed as Peak to Average Power Ratio.At some instant in time coherent addition of independently modulated N sub-carriers same phase produces a peak:

$$\text{PAPR}[x(t)] = \frac{P_{PEAK}}{P_{AVERAGE}}$$

$$= 10 \log_{10} \frac{\max[|X(n)|^2]}{E[|x_n|^2]} \quad (1)$$

Where P_{PEAK} represents peak output power, $P_{AVERAGE}$ means average output power. $E[\cdot]$ denotes the expected value, x_n represents the transmitted OFDM signals which are obtained by taking IFFT operation on modulated input symbols is expressed as:

$$x_n = \frac{1}{\sqrt{N}} \sum_{i=0}^{N-1} X_i e^{j \frac{2\pi}{N} ni} \quad (2)$$

Where N is the total number of subcarriers and the OFDM symbol duration is T seconds.

An OFDM symbol consists of N subcarriers by the frequency spacing of Δf . The total bandwidth B of a symbol will be divided into N equally spaced subcarriers and thus, all the subcarriers are orthogonal to each other within a time interval of length $T = 1/\Delta f$. A complex modulation symbol $X_{m,n}$, is used to modulate each sub-carrier independently where m is a time index and n is a subcarrier index. Then within the time interval T the following signal of the m -th OFDM block period can be described as:

$$x_m(t) = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} X_{m,n} b_n(t - mT) \quad (3)$$

Where, $b_n(t)$ is defined as:

$$b_n(t) = \begin{cases} \exp(j2\pi n \cdot f t), & 0 \leq t \leq T \\ 0, & \text{else} \end{cases} \quad (4)$$

Where, $b_n(t)$ is a rectangular pulse applied to each subcarrier. The total continuous time signal $x(t)$ consisting of all the OFDM block is given as:

$$x(t) = \frac{1}{\sqrt{N}} \sum_{m=0}^{\infty} \sum_{n=0}^{N-1} X_{m,n} g_n(t - mT) \quad (5)$$

When compared with traditional single carrier systems, the instantaneous output of an OFDM system often has large fluctuations. To handle these large fluctuations the system devices, such as power amplifiers, A/D converters and D/A converters, must have large linear dynamic ranges. If the peak signal goes into the non-linear region of devices at the transmitter, a series of undesirable interference is encountered, such as high out of band radiation and inter-modulation distortion. Therefore PAPR reduction techniques are required for OFDM systems. The power output has large fluctuations due to which, the high power amplifier (HPA) should have large dynamic range. This results in poor power efficiency.

OFDM systems combine a number of independent sub carriers which causes increase in PAPR. The increase in PAPR is related to the number of sub carriers and their order of modulation. A curve called as Complementary Cumulative Distribution Function (CCDF) is used to present the vital information regarding the OFDM signal to be transmitted. These curves also provide the PAPR data needed by component designer. The main use of power CCDF curves is to identify the power characteristic of the signals which are amplified, mixed and decoded [12]. The plot of relative power levels of signal against their probability of occurrence is called CCDF curve.

This curve illustrates the amount of time the signal remains at or above a given threshold level [13, 14]. The ratio between power level and the average power is expressed in dBas:

$$CCDF = \text{Prob}(PAPR > PAPR_o) \quad (6)$$

where $PAPR_o$ is the threshold level.

3. μ -LAW COMPANDING TECHNIQUE

In an OFDM system, large signals occur very infrequently therefore it is similar to the speech signal, so the same companding technique can be used to improve the OFDM transmission performance. One of the companding techniques named as μ -Law Companding can effectively reduce the Peak to Average Power Ratio (PAPR) of the transmitted or the companded OFDM signals by transforming the statistics of the amplitudes of these signals into uniform distribution. The key idea of the μ -law compander is that it covers all the encountered amplitudes of the OFDM signal. In μ -law companding, the uniform distribution of the signals can be obtained by compressing the peak signals and expanding the small signals. In the process of companding, the amplitudes of the small signals get enlarged, while the peaks remain unchanged. Therefore, the average power is increased and thus the Peak to Average Power Ratio (PAPR) can be reduced. This technique overcomes the deficiencies that are caused by clipping the original OFDM signal. The other advantage of the companding transform is that, it eliminates the Out of Band Interference (OBI) and can also maintain a constant average power level and also.

The original OFDM signal is converted into the companded signal by using the Mu Companding Transform. The companded signal obtained by using the Mu Companding Transform is given by the equation as:

$$H(x) = \frac{\text{sgn}(x) \ln(1 + \mu |x|)}{\ln(1 + \mu)} \quad (7)$$

Where, $H(x)$ = Companded Signal obtained by μ -Law Companding Technique,

$\text{sgn}(x)$ = sign function,

x = original OFDM signal,

$\mu = 255$ and it is the Mu-Law parameter of the compander, which controls the amount of compression.

4. DISCRETE COSINE TRANSFORM-II

The discrete cosine transform (DCT) is a Fourier-related transform similar to the discrete Fourier transform (DFT). It is a member of a family of sinusoidal unitary transforms. This transform is real, orthogonal and separable with fast algorithms for its computation. They have a better job of concentrating energy and compression of audio and data. The Discrete Cosine Transform (DCT) was first proposed by Ahmed *et al.* (1974) [15]. It is a Fourier-like transform, which is equivalent to a DFT of roughly twice the length, operating on real data with even symmetry. The discrete cosine transform-II is a linear, invertible function $F: R_n \rightarrow R_n$; where R denotes the set of real numbers, or equivalently an $n \times n$ square matrix. Then real numbers x_0, \dots, x_{n-1} are transformed into the n real numbers f_0, \dots, f_{n-1} according to one of the formulas:

$$f_j = \sum_{k=0}^{n-1} x_k \cos \left[\frac{\pi}{n} j \left(k + \frac{1}{2} \right) \right] \tag{8}$$

The DCT-II implies the boundary conditions: x_k is even around $k=-1/2$ and even around $k=n-1/2$; f_j is even around $j=0$ and odd around $j=n$.

The DCT II is important because it has better computational efficiency, superiority in bandwidth compression of a wide range of signals, its powerful performance in the bit-rate reduction and robust against inter carrier interference (ICI). The precoding based techniques are simple and linear techniques with less complexity to implement without the need of any side information.

5. PROPOSED SCHEME

A new technique named as precoding technique is also used in OFDM systems to reduce the PAPR. Precoding techniques obtain a signal with lower PAPR than in the case of a without precoded OFDM system and it also reduces the interference produced by multiple users. The μ -Law Companding Transform also causes less spectrum side-lobes as compared to other reduction techniques. In proposed scheme μ -Law companding was combined with precoding technique called Discrete Cosine Transform-II. The combined techniques based OFDM system was shown in fig. 2. In the transmitter, firstly DCT-II is used to transform the parallel data stream, and then the transformed data is processed by the companding unit. If data block passed by DCT-II, before IFFT, the PAPR is reduced, then the PAPR of OFDM signal could be further reduced by companding.

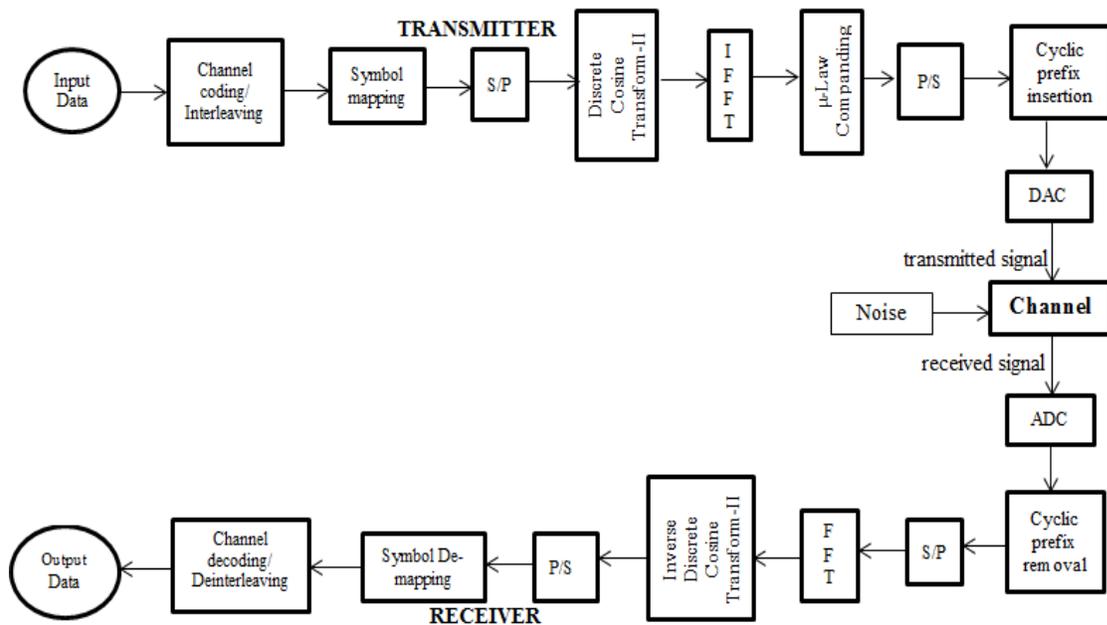


Fig. 2 Block scheme of DCT-II Precoding technique with μ -law companding in OFDM system

After applying the DCT-II on OFDM signal the signal get compressed decreasing the PAPR value. Mucompanding is performed on the I and Q outputs of the IFFT after precoding. As the word length at the IFFT output is decreased, the power consumption and complexity of the DAC/ADC decreases. Also companding at the IFFT output increases the resolution giving a better average signal.

6. SIMULATION RESULTS

To reduce the PAPR, the above discussed has been designed and results are shown in the form of CCDF vs. PAPR plot for different modulations and different values of M. In simulation, an OFDM system is considered under an additive white Gaussian noise (AWGN) channel, where data is randomly generated with subcarriers $N=2400$, then the signal is modulated by M -PSK and M -QAM respectively. The PAPR reduction capability is measured by the complementary cumulative distribution ($CCDF = Prob (PAPR > PAPR_0)$), which indicates the probability that PAPR is above a certain threshold. We compared the simulation results of proposed system with DCT-II precoded OFDM signal and precoded μ -Law companded signal. The results were observed for precoded signal and for precoded μ -Law companded signal.

6.1 M-PSK Modulation

In this section μ -Law companding technique is applied over DCT-II precoded OFDM signal with phase shift keying modulation. We assume CCDF clip rate of $10^{-0.6}$ using M-ary PSK technique for $N=2400$ and $M=16, 32, 64, 128, 256, 512$ and 1024 .

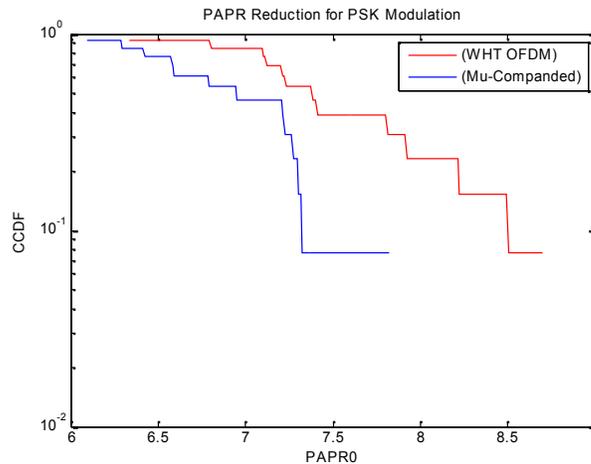


Fig. 3: CCDF of μ -Law companding with proposed DCT-II technique for 16 PSK

For OFDM system CCDF vs. PAPR plots using $M=16$ over AWGN channel is shown in Fig. 3. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.53 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

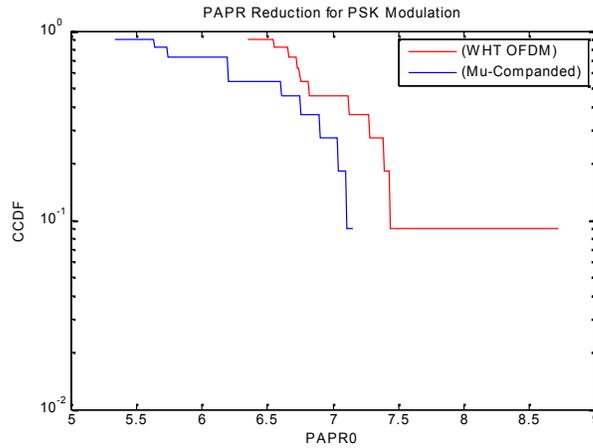


Fig. 4: CCDF of μ -Law companding with proposed DCT-II technique for 32 PSK

For OFDM system CCDF vs. PAPR plots using $M=32$ over AWGN channel is shown in Fig. 4. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.51 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

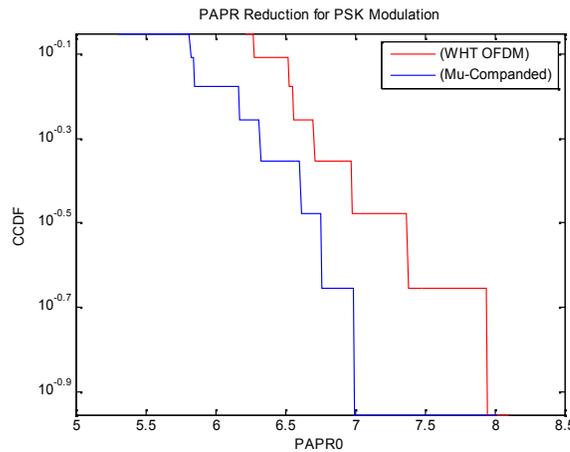


Fig. 5: CCDF of μ -Law companding with proposed DCT-II technique for 64 PSK

For OFDM system CCDF vs. PAPR plots using $M=64$ over AWGN channel is shown in Fig. 5. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.7 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

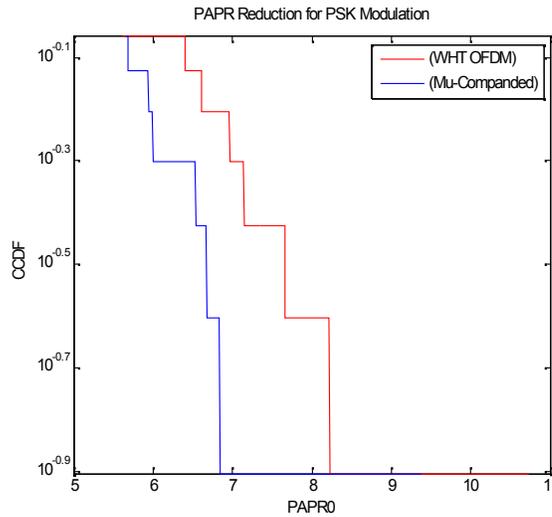


Fig. 6: CCDF of μ -Law companding with proposed DCT-II technique for 128 PSK

For OFDM system CCDF vs. PAPR plots using $M=128$ over AWGN channel is shown in Fig. 6. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 1.03 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

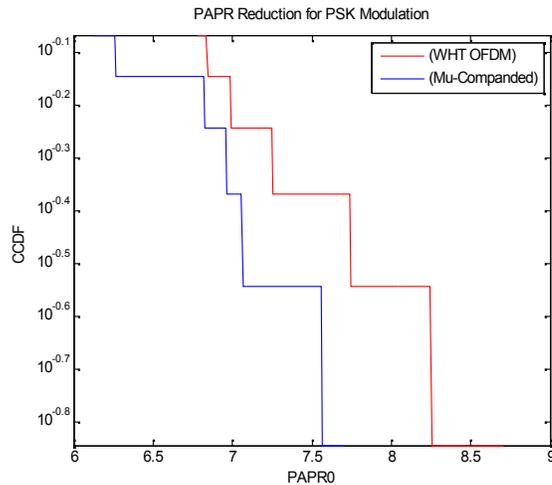


Fig. 7: CCDF of μ -Law companding with proposed DCT-II technique for 256 PSK

For OFDM system CCDF vs. PAPR plots using $M=256$ over AWGN channel is shown in Fig. 7. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.65 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

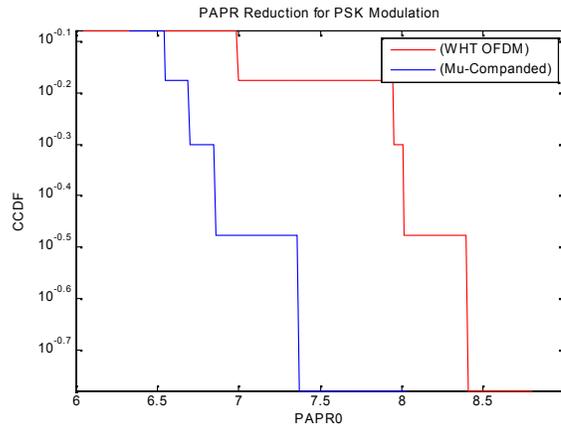


Fig. 8: CCDF of μ -Law companding with proposed DCT-II technique for 512 PSK

For OFDM system CCDF vs. PAPR plots using $M=512$ over AWGN channel is shown in Fig. 8. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 1.02 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

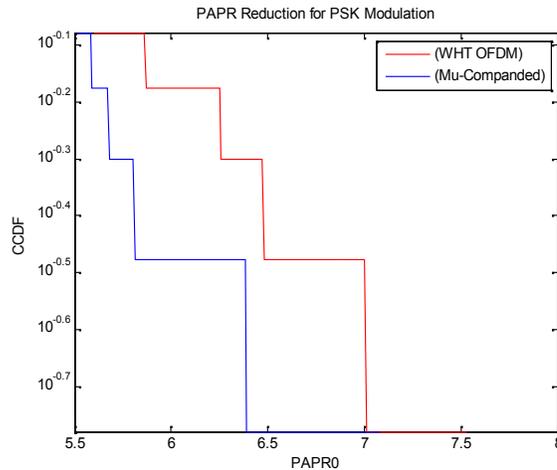


Fig. 9: CCDF of μ -Law companding with proposed DCT-II technique for 1024 PSK

For OFDM system CCDF vs. PAPR plots using $M=1024$ over AWGN channel is shown in Fig. 9. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.62 dB using PSK modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

6.2 M-QAM Modulation

For $M=16, 64, 256$ and 1024 M-ary quadrature amplitude modulation (QAM) technique is used here at CCDF rate of $10^{-0.6}$.

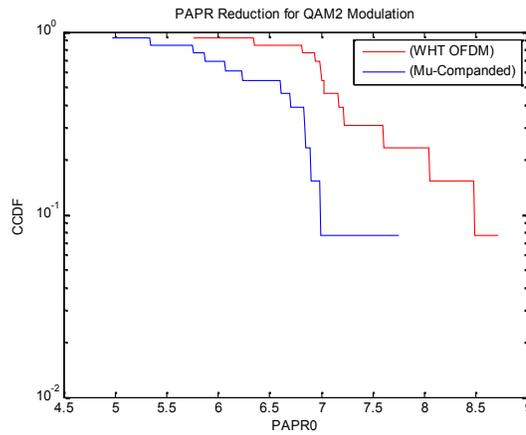


Fig. 10: CCDF of μ -Law companding with proposed DCT-II technique for 16 QAM

For OFDM system CCDF vs. PAPR plots using $M=16$ over AWGN channel is shown in Fig. 10. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.87 dB using QAM modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

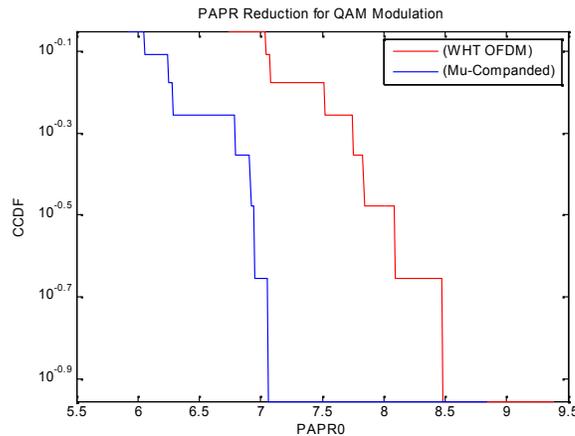


Fig. 11: CCDF of μ -Law companding with proposed DCT-II technique for 64 QAM

For OFDM system CCDF vs. PAPR plots using $M=64$ over AWGN channel is shown in Fig. 11. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 0.83 dB using QAM modulation. The proposed μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

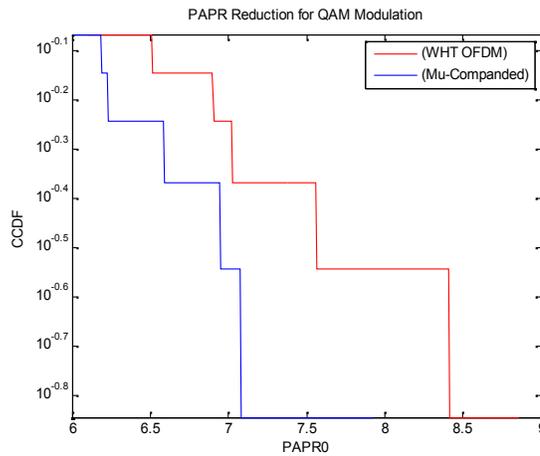


Fig. 12: CCDF of μ -Law companding with proposed DCT-II technique for 256 QAM

For OFDM system CCDF vs. PAPR plots using $M=256$ over AWGN channel is shown in Fig. 12. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 1.23 dB using QAM modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

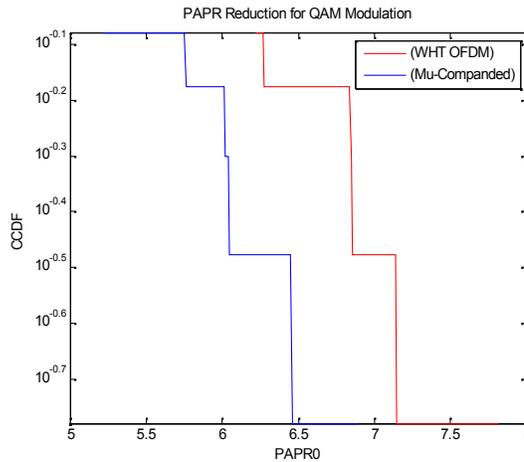


Fig. 13: CCDF of μ -Law companding with proposed DCT-II technique for 1024 QAM

For OFDM system CCDF vs. PAPR plots using $M=1024$ over AWGN channel is shown in Fig. 13. The graph gives the clear idea that at CCDF rate of $10^{-0.6}$ the PAPR value decreases by 1.3 dB using QAM modulation. The proposed precoded μ -Law companded technique provides much reduction in PAPR value as compared to DCT-II precoded signal.

7. CONCLUSIONS

In this paper, a new idea about the performance of the μ -Law companding technique combined with precoded technique is presented for the reduction of PAPR value. The PAPR value can be reduced for different modulation formats and for different values of M by using proposed

scheme without increasing the system complexity. The scheme shows superiority in bandwidth compression of signals and powerful performance in the bit-rate reduction. This system produce the lowest PAPR and is efficient, signal independent, distortion less and does not causes any spectrum side-lobes. The simulation results shows, that the proposed precoded combined μ -law companding method displays a better PAPR reduction performance than the simple DCT-II precoded OFDM signal. Thus, it is concluded that the proposed scheme reduces PAPR effectively compared with precoded DCT-II transform.

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